

# POWER OPERATIONAL AMPLIFIER

# PAD108

Rev D

# **KEY FEATURES**

- LOW COST
- HIGH VOLTAGE 200 VOLTS
- HIGH CURRENT 10 AMPS
- 100 WATT DISSIPATION CAPABILITY
- 300kHz POWER BANDWIDTH

# **APPLICATIONS**

- INKJET PRINTER HEAD DRIVE
- PIEZO TRANSDUCER DRIVE
- INDUSTRIAL INSTRUMENTATION
- **RELECTOMETERS**
- ULTRA-SOUND TRANSDUCER DRIVE

## **DESCRIPTION**

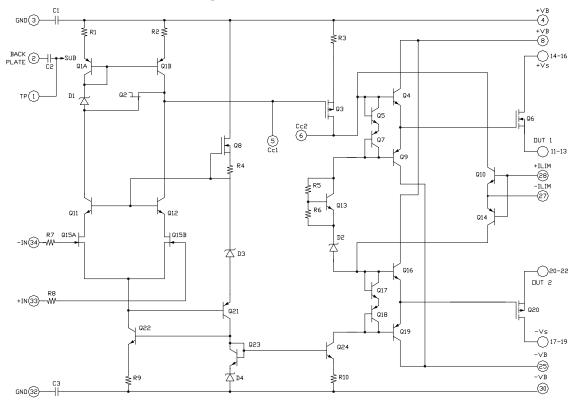
The PAD108 is a high voltage operational surface amplifier constructed with mount components to provide a cost effective solution for many industrial applications With a footprint only 3.3in<sup>2</sup> the PAD108 offers outstanding performance that rivals more expensive hybrid component amplifiers. User selectable external compensation tailors the amplifier's response to the application requirements. Four-wire current limit is built-in. The amplifier circuitry is built on a thermally conductive but electrically insulating substrate. No BeO is used in the PAD108. The resulting module is a small, high performance solution for many industrial applications.



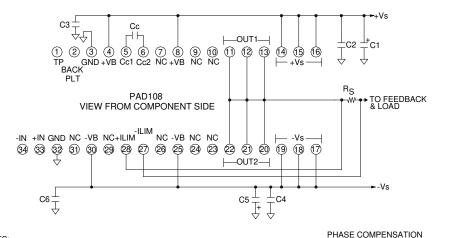




EQUIVALENT CIRCUIT



## **PINOUT & CONNECTIONS**



NOTES.			
1. Cc IS NPO (COG) RATED FOR FULL SUPPLY VOLTAGE +Vs TO -Vs	GAIN W/O BOOST	<u>Cc</u> 100pF	SLEW RATE 55V/uS
2. BOTH PINS 3 AND 32 ARE REQUIRED CONNECTED TO SIGNAL GROUND	4	33pF	135V/uS
3. C1 AND C5 ELECTROLYTIC, 10uF PER AMP OUTPUT CURRENT	≥ 10	10pF	170V/uS
4. C2,3,4,6 HIGH QUALITY CERAMIC 0.1uF			
5, ALL OUTPUT PINS MUST BE TIED TOGETHER	GAIN W BOOST	Cc	
	1	470pF	12V/uS
	3	220pF	35V/uS
	≥ 10	33pF	135V/uS







### ABSOLUTE MAXIMUM RATINGS **SPECIFICATIONS**

### ABSOLUTE MAXIMUM RATINGS

SUPPLY VOLTAGE, +Vs to -Vs	200
BOOST VOLTAGE,	±Vs
OUTPUT CURRENT, within SOA	12A
POWER DISSIPATION, internal, DC	100
INPUT VOLTAGE, differential	± 20
INPUT VOLTAGE, common mode	$\pm V_{E}$

)V s ±20V A )W 0V В

TEMPERATURE, pin solder, 10s TEMPERATURE, junction<sup>2</sup> **TEMPERATURE RANGE**, storage **OPERATING TEMPERATURE**, case

237C 150°C -40 to 105°C -40 to 85°C

PARAMETER	TEST CONDITIONS <sup>1</sup>	MIN	ТҮР	MAX	UNITS
INPUT					
OFFSET VOLTAGE			1	3	mV
OFFSET VOLTAGE vs. temperature	Full temperature range		20	50	μV/ <sup>o</sup> C
OFFSET VOLTAGE vs. supply				20	μV/V
BIAS CURRENT, initial <sup>3</sup>				100	pA
BIAS CURRENT vs. supply				0.1	pA/V
OFFSET CURRENT, initial				50	pA
INPUT RESISTANCE, DC			100		GΩ
INPUT CAPACITANCE			4		pF
COMMON MODE VOLTAGE RANGE				$+V_{B}-15$	V
COMMON MODE VOLTAGE RANGE				-V <sub>B</sub> +7	V
COMMON MODE REJECTION, DC		98	106		dB
NOISE	100kHz bandwidth, $1k\Omega R_s$		10		μV RMS
GAIN					
OPEN LOOP	$R_L = 10k\Omega, C_C = 10pF$	108			dB
GAIN BANDWIDTH PRODUCT @ 1MHz	$C_{\rm C}=10\rm{pF}$		10		MHz
PHASE MARGIN	Full temperature range	45	60		degree
OUTPUT					
VOLTAGE SWING	$I_0 = 10A$	±Vs-10	+Vs-8.6		V
VOLTAGE SWING	$I_{\Omega} = -10A$	-Vs+10	-Vs+7		V
VOLTAGE SWING	$+V_{B}=+V_{S}+10V, I_{O}=10A$	+Vs-1.6	+Vs-1.2		
VOLTAGE SWING	$-V_{B}=-V_{S}-10V, I_{O}=-10A$	-Vs-5.1	+Vs+4.1		
CURRENT, continuous, DC		11			Α
SLEW RATE, $A_V = -10$	$C_{\rm C} = 10 \rm pF$	150	170		V/µS
SETTLING TIME, to 0.1%	$2V$ Step, $C_C = 10 pF$		1		μS
RESISTANCE	No load, DC		4		Ω
POWER BANDWIDTH, 180Vp-p	$C_{\rm C} = 10 \rm pF$		300		kHz
POWER SUPPLY					
VOLTAGE		±15	±75	±100	V
CURRENT, quiescent			50	65	mA
THERMAL					
RESISTANCE, AC, junction to case <sup>5</sup>	Full temperature range, $f \ge 60$ Hz			1	<sup>o</sup> C/W
RESISTANCE, DC junction to case	Full temperature range			1.25	°C/W
RESISTANCE, DC junction to case	Full temperature range			1.25	°C/W
TEMPERATURE RANGE, case		-40		85	°C

NOTES:

1. Unless otherwise noted:  $T_c = 25^{\circ}$ C, compensation Cc = 100pF, DC input specifications are  $\pm$  value given, power supply voltage is typical rating.

2. Derate internal power dissipation to achieve high MTBF.

3. Doubles for every 10°C of case temperature increase.

4. +Vs and -Vs denote the positive and negative supply voltages to the output stage. +V<sub>B</sub> and -V<sub>B</sub> denote the positive and negative supply voltages to the input stages.

5. Rating applies if the output current alternates between both output transistors at a rate faster than 60Hz.

7. Power supply voltages  $+V_{B}$  and  $-V_{B}$  must not be less than  $+V_{S}$  and  $-V_{S}$  respectively. Total voltage  $+V_{B}$  to  $-V_{B}$  240V maximum.

8. The PAD108 is constructed with MOSFET transistors and ESD handling procedures must be observed.





### **SAFETY FIRST**

The operating voltages of the PAD108 are potentially deadly. When developing an application circuit it is wise to begin with power supply voltages as low as possible while checking for circuit functionality. Increase supply voltages slowly as confidence in the application circuit increases. Always use a "hands off" method whereby test equipment probes are attached only when power is off.

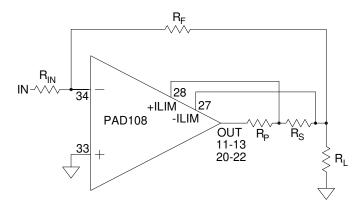
### **CURRENT LIMIT**

The current limiting function of the PAD108 is a versatile circuit that can be used to implement a four-wire current limit configuration. The four-wire current limit configuration insures that parasitic resistance in the output line, Rp, does not affect the programmed current limit setting. See Figure 1 below. The sense voltage for current limit is 0.7V. Thus:

$$I_L = \frac{0.65V}{R_s}$$

Where  $I_L$  is the value of the limited current and  $R_S$  is the value of the current limit sense resistor.

In addition, the sense voltage has a temperature coefficient approximately equal to -2.2 mV/°C.



#### Figure 1

#### **MOUNTING THE AMPLIFIER**

In most applications the amplifier must be attached to a heat sink. Spread a thin and even coat of heat sink grease across the back of the PAD108 and also the heat sink where the amplifier is to be mounted. Push the amplifier into the heat sink grease on the heat sink while slightly twisting the amplifier back and forth a few times to bed the amplifier into the heat sink grease. On the final twist align the mounting holes of the amplifier with the mounting holes in the heat sink and finish the mounting using 4-40 hex male-female spacers. Mount the amplifier to the mother board with 4-40 X 1/4" screws.

#### PHASE COMPENSATION

The PAD108 **must** be phase compensated. The compensation capacitor,  $C_c$ , is connected between pins 5 and 6. The compensation capacitor must be an NPO type capacitor rated for the full supply voltage (200V). On page 2, under Amplifier Pinout and Connections, you will find a table that gives recommended compensation capacitance value for various circuit gains and the resulting slew rate for each capacitor value. Consult also the small signal response and phase response plots for the selected compensation value in the Typical Performance Graphs section Do not use a compensation capacitor less than 10pF.

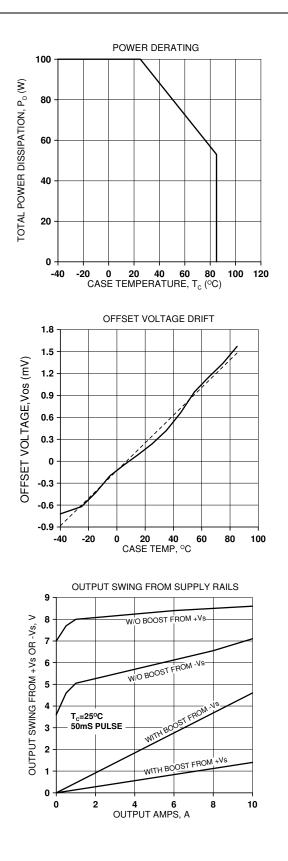
#### **BOOST OPERATION**

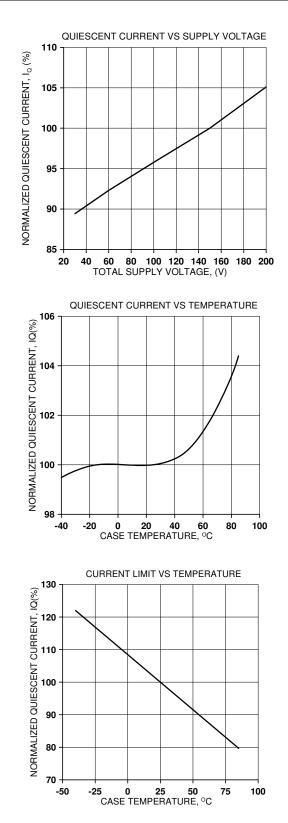
The small signal stages of the PAD108 are connected to the  $\pm V_B$  power supply pins. When the  $\pm V_B$  voltages are greater than the  $\pm V_S$  power supply pins the small signal stages of the amplifier are biased so that the output transistors can be driven very close to the  $\pm V_S$  rails. Close swings to the supply rails increase the efficiency of the amplifier and make better use the supply voltages. This technique is often used to operate the amplifier with only a single high current power supply, thus reducing the system size and cost. Also see the application article AN-22 *Single Supply Operation with Power Op Amps* for more detailed information and circuits.





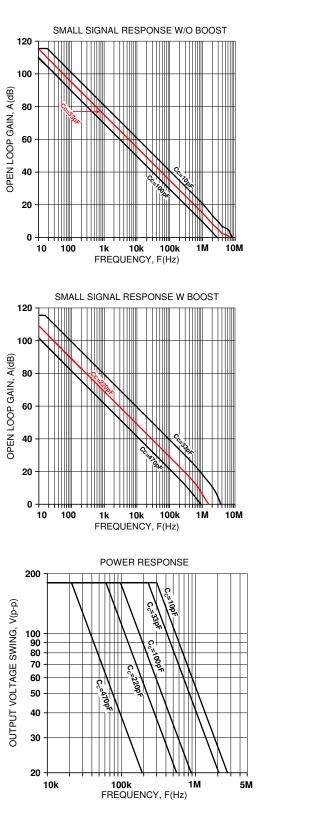
# PAD108 TYPICAL PERFORMANCE GRAPHS

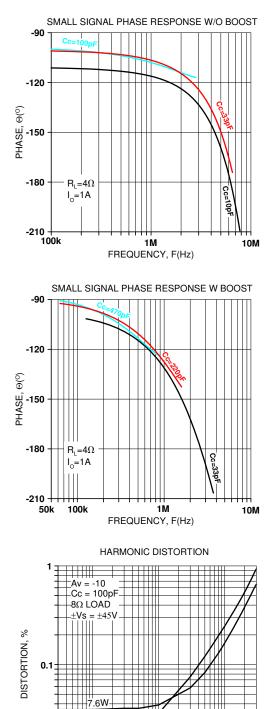














76V

1k

FREQUENCY, F(Hz)

100

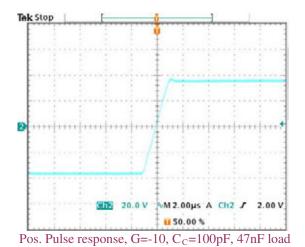
0.01

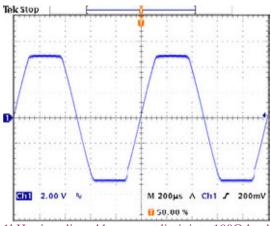
30

30k

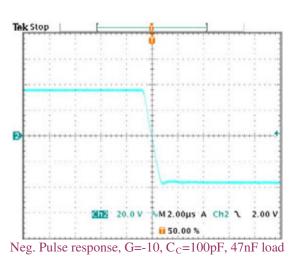
10k

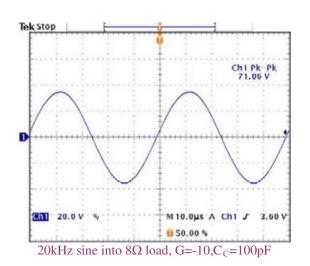






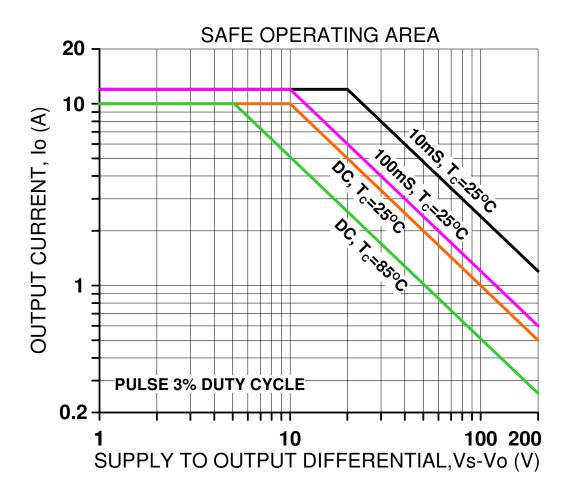
1kHz sine clipped by current limit into  $100\Omega$  load











SAFE OPERATING AREA





